

Vehicle Warning System using Dynamic Pattern Projection with Active Mirror on Road to avoid collision between Vehicle and Pedestrians.

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ABSTRACT: Recently, the demand for vehicle warning systems to avoid collisions between vehicles and pedestrians has been increasingly growing. For the implementation of such systems, three types of vehicle-to-pedestrian (V2P) technologies are widely adopted. In the first, vehicles communicate with pedestrians via smart devices using dedicated short-range communications (DSRC), but it is neither simply intuitive nor fast responsive. In the second, vehicles generate a noise in order to help pedestrians avoid collisions, at the expense of disturbing others with unwelcome noise. In the last, an on-road short-range projection technique visually informs pedestrians that a vehicle is approaching, which can be more intuitive and responsive. However, the two main drawbacks of conventional on-road projections are that they involve a static pattern and narrow emission angle. This paper proposes an on-road dynamic pattern projection technique for grabbing more attention from pedestrians with various dynamic patterns and covering a wide emission angle using an active mirror. Theoretical analysis shows that the on-road projection pattern and its angle can easily be controlled by adjusting the switching frequency, duty-cycle, and amount of current flow in the control circuitry. The simulation and experimental results validate that various patterns can be projected within a projection angle of $\pm 70^\circ$ at a power consumption less than 0.98W.

KEYWORDS: Vehicle warning, vehicle-to-pedestrian (V2P), on-road projection, dynamic pattern, active mirror.

I. INTRODUCTION

GOVERNORS highway safety association (GHSA) announced that the number of pedestrian fatalities increased from 2007 to 2016 by 27% [1].

In addition, the national highway traffic safety administration (NHTSA) reported in 2016 that there were 5,987 pedestrian fatalities in traffic crashes in the United States [2]. It is important to note that the reports by both GHSA and NHTSA indicated that 75% of pedestrian fatalities occurred in the dark. Furthermore, NHTSA reported that an average of 210 fatalities and 15,000 injuries per year are caused by back over crashes. Not-in-traffic surveillance (NiTS) system reported similarly that 284 fatalities and 12,000 injuries on average occur every year due to back over crashes [3]. Children aged five and under account for 31% of back over fatalities each year, while elderly people aged 70 and over account for 26%. In order to prevent these accidents by warning pedestrians, several vehicle-to-pedestrian (V2P) technologies have emerged. First, dedicated short-range communications (DSRCs) have been developed by forwarding warning signals to pedestrians via smart devices [4]–[6], but these are not useful for people without smart devices, particularly vulnerable road users (VRUs) such as children and elderly people who do not have a smart device or do not know how to use it. Second, a back-up beeper can be used by generating sound to warn pedestrians about an impending collision, but this can disturb people who are not the target of the beeper [7]–[9]. Unlike the other two techniques, on-road projection is intuitive and has been widely researched in application to a limited area [10]–[16]. As shown in Fig. 1, a light pattern is projected onto the road so as to warn pedestrians that a vehicle is approaching or around them. The on-road projection is a means to provide pedestrians with an intuitive sense of the vehicle, even if they are not carrying smartphones.



Fig. 1. Conventional vehicle warning system using on-road short-range projection with a static pattern.

As for the implementation of the on-road projection, two methods can be used: using collimating lens or digital micro mirror devices (DMDs). Using a collimating lens requires a static patterned film [10], [11], but the light emission angle is proportioned to the size of the lens, resulting in a large volume. The other is a way to project pixel images through the digital micro-

mirror device [12]–[14]. The drawbacks of this system are the limited emission angle and the usage of customized optical components that raises the cost of the system. In this paper, we propose an efficient on-road projection system which can project various dynamic patterns with a wide emission angle and does not require customized optics.

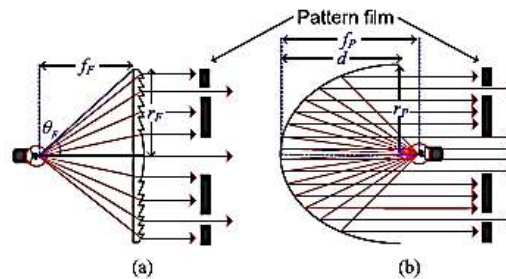


Fig. 2. Collimating lens for on-road projection (a) Fresnel lens, (b) parabolic reflector.

II. BASIC IDEA

A. Conventional Techniques

Conventional on-road projection techniques with a collimating lens, as shown in Fig. 2, are composed of a light source, a collimator, and a pattern film. Fig. 2(a) shows a system with a Fresnel lens of which radius r_f is equal to $f_f \tan \theta_f$, where f_f is the focal length and θ_f is the emission angle of the light source [11]. To have a larger pattern, larger volume cannot be avoided due to increased values of either f_f or θ_f . The radius r_p of the parabolic reflector shown in Fig. 2(b) is proportional to the square root of $f_p d$, where f_p is the focal length and d is the depth of the parabolic reflector [12]. There are several drawbacks to these techniques. First, the size of the optics for collimating light increases in order to allow for the projected pattern to be enlarged, since the radius of the projected patterns for both techniques is proportional to their focal length. Second, the projected light cannot be dynamically changed, resulting in a fixed pattern that may fail to attract pedestrians [10], [11]. Another technique for on-road projection adopts DMDs with projection lens [12]–[14]. While it is advantageous to have

dynamic patterns that can be projected using numerous pixels, its patterns cannot be projected ideally due to the fact that the tilting angle of DMDs is limited to $\pm 15^\circ$. Moreover, the on-road projection with DMDs requires customized optics for the alignment of the lights reflected by the micro-mirror array, which raises the cost of the system.

B. Proposed Technique

Fig. 3 shows the concept of the proposed dynamic on-road projection with an active mirror and multiple laser diodes (LDs). The lights from the fixed LDs are reflected from the mirror and projected onto the road. Then, various dynamic patterns can be formed by controlling the tilting angle θ of the active mirror. The proposed system is composed of two main blocks: LD and

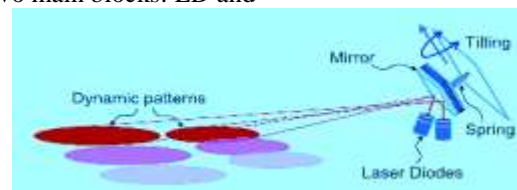


Fig. 3. Proposed dynamic on-road projection using an active mirror and multiple laser diodes.

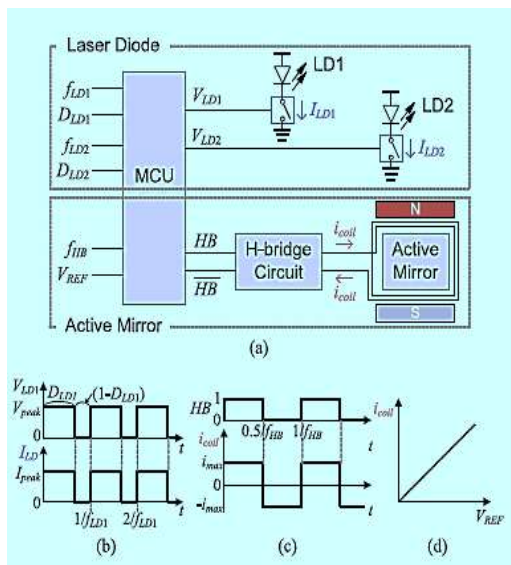


Fig. 4. Proposed system (a) block diagram, (b) timing diagram for the laser diode, (c) timing diagram for the active mirror, (d) V_{REF} vs. i_{coil} .

Active mirror. In the LD block, a micro-controller unit (MCU) controls the switching frequency f_{LD} and the duty-cycle D_{LD} of LDs, as shown in Fig. 4(a). Then, the current through the LDs, IPEAK, can be controlled as shown in Fig. 4(b). In the active mirror block, the MCU controls the switching frequency of H-bridge, f_{HB} , with the duty-cycle of 0.5 and the current through the coil, i_{coil} , by adjusting V_{REF} as shown in Figs. 4(c) and 4(d). The tilting angle of the active mirror is then the function of f_{HB} and i_{coil} , and its relationship will be described in the next section.

The basic structure of the proposed active mirror consists of two permanent magnets, a spring, and a mirror with a coil, as shown in Fig. 5. The mirror is mounted on the fixed frame via the spring and located in between two permanent magnets with a constant magnetic field. The current through the coil and the magnetic field allows for the mirror to tilt in a given direction. If a current is supplied to the coils in one direction as shown in Fig. 6(a), Lorentz forces $\mathbf{FP1}$ and $\mathbf{FP2}$ are applied to two points $\mathbf{P1}$ and $\mathbf{P2}$ located on each side in opposing directions and let the mirror tilt in one direction by the moment. The higher number of turns in a coil or the larger the current, the larger the net current

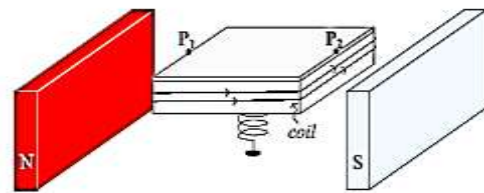


Fig. 5. Structure of the proposed active mirror.

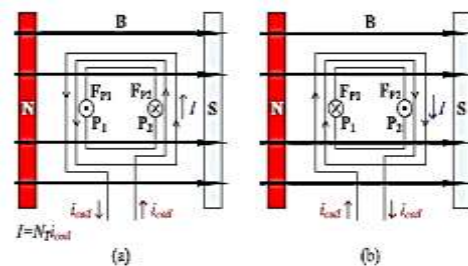


Fig. 6. Top view of the proposed active mirror. Lorentz forces applied at $\mathbf{P1}$ and $\mathbf{P2}$ when (a) the current in the coil flows counter-clockwise and when (b) the current in the coil flows clockwise.

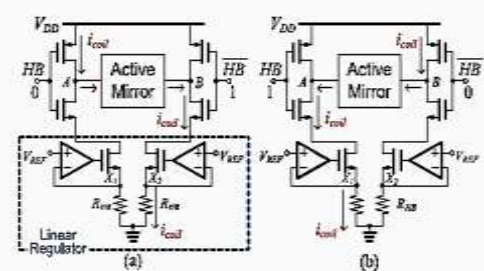


Fig. 7. Reversion of the coil current's direction depending upon whether the digital input HB is (a) low or (b) high.

I and the greater the Lorentz force \mathbf{FL} is applied as $\mathbf{FL} = \mathbf{FP1} = -\mathbf{FP2} = \mathbf{B}n\text{icoill} = \mathbf{B}lI$, (1) where \mathbf{B} is the magnetic field, n the number of turns, i_{coil} the current through coil, and l the length of the wire. Once the current flow is reversed, the mirror plane tilts in the opposite direction, as depicted in Fig. 6(b) by the Lorentz forces applied in opposing directions at $\mathbf{P1}$ and $\mathbf{P2}$. The main advantage of the proposed active mirror, as compared to a mirror controlled by a DC motor, is that the resilience of the spring allows for automatic alignment of the active mirror without the need for an additional control to check the tilting angle of the mirror. In order to control the direction of the current through the coil and the amount of current, an H-bridge with an active mirror is proposed, as depicted in Fig. 7. Similar to the conventional H-bridge circuits presented in [17]–[23], the control HB determines the

direction of the current flowing through the active mirror. If HB is low, as shown in Fig. 7(a), the current flows from node A to B. Once HB is set to a high value, as in Fig. 7(b), the current flows from the node B to A. In addition, the amount of the current icoil can be adjusted to the desired value through the linear regulator, and can be defined as $icoil = VREF/RHB$. (2)

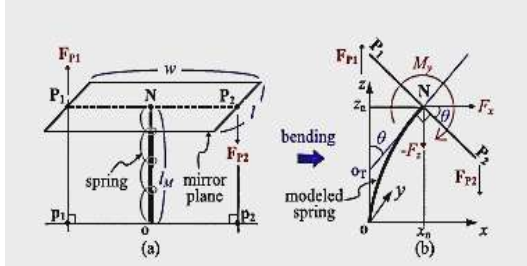


Fig. 8. Static analysis of the proposed active mirror (a) angled view of the active mirror without bending forces, (b) simplified cross-sectional view of the active mirror with bending forces.

III. MATHEMATICAL MODELING

A. Force and Moment Balance Equations

Fig. 8(a) shows an angled view of the proposed active mirror. The mirror is simplified as a plane, and the spring is an elastic bar, since the lateral buckling of a coil compression spring can be treated the same as an elastic bar [28]. The static analysis for the moment and net forces in the x-, y-, and z-axis is based on the assumption that the moment and all of the forces acting on the active mirror can be transformed to the moment and the forces at the center of the active mirror plane [24]–[27]. The mirror plane with a tilting angle θ about axis z, as shown in Fig. 8(b), can be obtained using the rotational matrix \mathbf{R} from Rodrigues' formula in [24] as $\mathbf{R} = \mathbf{I} + \hat{\omega} \sin \theta + \hat{\omega}^2 (1 - \cos \theta)$

$$= \begin{bmatrix} \cos \theta & 0 & \sin \theta \\ 0 & 1 & 0 \\ \sin \theta & 0 & \cos \theta \end{bmatrix} \quad (3)$$

$$\mathbf{T} = \begin{bmatrix} \cos \theta & 0 & \sin \theta & x_n \\ 0 & 1 & 0 & 0 \\ -\sin \theta & 0 & \cos \theta & z_n \\ 0 & 0 & 0 & 1 \end{bmatrix} \quad (4)$$

Using the homogeneous transformation matrix in Eq. (4), as shown in Fig. 8(b), new vector coordinates of the mirror plane at location P1 and P2 after bending can be expressed as

$$\mathbf{P1} = \mathbf{T} \cdot \mathbf{p1}, \mathbf{P2} = \mathbf{T} \cdot \mathbf{p2}, \mathbf{N} = \mathbf{T} \cdot \mathbf{o}, \quad (5)$$

Defining w as the width of the mirror plane, then, each position can be addressed as

$$\mathbf{p1} = [-w/2, 0, 0, 1]^T, \mathbf{p2} = [w/2, 0, 0, 1]^T,$$

$$\mathbf{o} = [0, 0, 0, 1]^T.$$

Since the Lorentz forces acting on the active mirror are always perpendicular to the ground, they can be defined as

$$\mathbf{FP1} = [0, 0, FL]^T, \mathbf{FP2} = [0, 0, -FL]^T. \quad (6)$$

where FL is the scalar of Lorentz force \mathbf{FL} . Then, the moment and force balance equations of the proposed system can be

$$[\mathbf{Mx}, \mathbf{My}, \mathbf{Mz}]^T = \overline{\mathbf{NP1}} \times \mathbf{FP1} + \overline{\mathbf{NP2}} \times \mathbf{FP2}, \quad (7)$$

$$[\mathbf{Fx}, \mathbf{Fy}, \mathbf{Fz}]^T = \mathbf{FP1} + \mathbf{FP2} + [0, 0, -mg]^T. \quad (8)$$

where Mx, My, Mz and Fx, Fy, Fz are the scalar moments and forces acting on the active mirror at the center of the mirror. Then they can be obtained as

$$Mx = 0, My = wFL \cos \theta, Mz = 0,$$

$$Fx = 0, Fy = 0, Fz = -mg. \quad (9)$$

B. Origin of Tilting Axis

Since the change in the length of the spring by the weight of the mirror plane is not negligible, the change in length of the spring can be obtained using Hooke's law as

$$Fz = -mg = -k (l_0 - l_m), \quad (10)$$

where m is the weight of the mirror plane, k the spring constant, and l_0 the initial length of the spring without locating the mirror on top [24], [26]. The spring can be taken into account as a flexible bar in order to investigate the lateral buckling characteristics of the spring [24], [28]. Considering the situation shown in Fig. 8, the spring will be buckled by moment My and force $-Fz$. For any cross-section of the spring, the equation for its deformation can be defined as

$$EI_{sm} \frac{d^2 x/dz^2}{[1 + (dx/dz)^2]^{3/2}} = My - Fz (x_n - x). \quad (11)$$

where E is the elastic modulus and I_{sm} is the second moment of inertia [24], [28]. The right-hand side is the total moment applied to a cross-section of the spring. This equation does not have an analytical solution, but the numerical solution can be obtained by calculating elliptic integrals [29]. For small deflection, Eq. (11) can be linearized as

$$EI_{sm} \frac{d^2 x}{dz^2} = My - Fz (x_n - x). \quad (12)$$

The two initial conditions for two functions $x(z)$ and dx/dz for $z = 0$ are $x(0) = 0$ and $dx/dz = 0$, which leads to the solution as

$$x = -\frac{M_y - F_z x_n}{F_z} \left[1 - \cos \left(\sqrt{\frac{-F_z}{EI_{sm}}} z \right) \right]. \quad (13)$$

Then, the tangent equation at $z = z_n$ becomes

$$x = -\frac{M_y - F_z x_n}{F_z} \sqrt{\frac{-F_z}{EI_{sm}}} \sin \left(\sqrt{\frac{-F_z}{EI_{sm}}} z_n \right) (z - z_n) - \frac{M_y - F_z x_n}{F_z} \left[1 - \cos \left(\sqrt{\frac{-F_z}{EI_{sm}}} z_n \right) \right]. \quad (14)$$

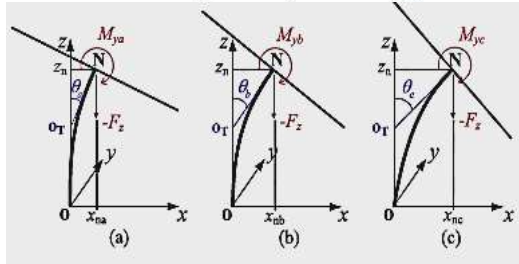


Fig. 9. Change in θ depending on different values of moment M_y (a) $M_y = M_{ya}$, (b) $M_y = M_{yb}$, (c) $M_y = M_{yc}$, where $M_{ya} < M_{yb} < M_{yc}$.

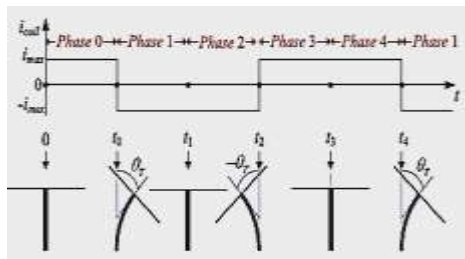


Fig. 10. Movement of the proposed active mirror depending on time.

Letting the crossing point between the tangent line and z -axis be the point oT , then the line between origin o and oT can be obtained as

$$\overline{ooT} = z_n - \frac{\left[1 - \cos \left(\sqrt{\frac{-F_z}{EI_{sm}}} z_n \right) \right]}{\sqrt{\frac{-F_z}{EI_{sm}}} \sin \left(\sqrt{\frac{-F_z}{EI_{sm}}} z_n \right)}. \quad (15)$$

Since the line length is a constant regardless of the moment as shown in Fig. 9, the tilting angle θ becomes $\theta = \angle noT zn$. Thus, oT can be defined as the origin of the tilting axis.

C. Movement of the Active Mirror

Fig. 10 describes the movement of the proposed active mirror depending on time. There are five different phases, and after phase 0, phases 1 to 4 are repeated periodically. The phase 0 is the initial phase with which to set the target tilting angle θT ,

where the initial velocity of the active mirror is zero. In both phases 1 and 3, θ returns to 0° from $|\theta T|$ after $t_1 - t_0 = t_3 - t_2$ with the non-zero initial velocity of the mirror. Similarly, in phases 2 and 4, θ becomes $|\theta T|$ after some time $t_2 - t_1 = t_4 - t_3$. The operation mechanism of the proposed active mirror for the angle of $\pm\theta T$ is as follows. In phase 0, the current through the coil, i_{coil} , is applied at $t=0$. At the beginning of phase 1 ($t=t_0$), the current through the coil becomes $-i_{coil}$ and keeps its value throughout the phase. Then, the direction of the current through the coil reverses every $t = t_0 + nN(t_2 - t_0)$, where nN is the natural number. Thus, the active mirror tilts periodically while maintaining the tilting angle of $\pm\theta T$, as shown in Fig. 10. As a result, the switching frequency of the H-bridge is determined as

$$f_{HB} = \frac{1}{2(t_2 - t_0)} = \frac{1}{t_4 - t_0}. \quad (16)$$

TABLE I
TERMS USED IN KINETIC DIAGRAMS

Symbol	QUANTITY	Symbol	QUANTITY
FL	Lorentz force	kb	lateral bending stiffness
B	magnetic flux density	θ	tilting angle of mirror
n	turns of coil	I_m	moment of inertia
I_{coil}	current flowing in a wire	M_r	moment by Lorentz force
I	product of n and I_{coil}	M_c	moment by gravity
/	length of wire receiving	M_k	moment by spring's resilience

	Lorentz force		
m	mass of moving plate	Maet	net moment of active mirror
g	acceleration of gravity		

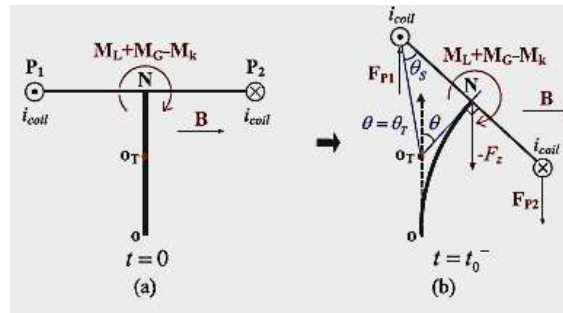


Fig. 11. Movement of the active mirror from the time (a) $t = 0$ to (b) $t = t_0$.

where $t_1 - t_0 = t_3 - t_2$ and $t_2 - t_1 = t_4 - t_3$. In order to obtain fHB, dynamic analysis of the active mirror is required. The dynamic analysis for each phase based on the parameters in Table I is as follows.

1) Phase 0: Lorentz forces are applied to both sides of the mirror plane with the current i_{coil} to the coil at the range of $0 \leq t \leq t_0$ as shown in Fig. 11. Then moments MP_1 and MP_2 at point oT by Lorentz forces FP_1 and FP_2 , respectively, can be defined as

$$M_{P_1} = \vec{oT} \times \vec{P_1} \times F_{P_1} = \left[0, \frac{w}{2} \sec \theta_s F_L \sin(90^\circ - \theta_s - \theta), 0 \right], \quad (17)$$

$$M_{P_2} = \vec{oT} \times \vec{P_2} \times F_{P_2} = \left[0, \frac{w}{2} \sec \theta_s F_L \sin(90^\circ - \theta_s - \theta), 0 \right]. \quad (18)$$

By the superposition of the moments at point oT , ML , all of the Lorentz forces can be expressed as

$$ML = MP_1 + MP_2 = [0, wFL \cos \theta, 0], \quad (19)$$

Since the scalar quantity of the y-axis component is positive, the tilting of the mirror plane is in a clockwise direction. Another moment at point oT by the gravity and the mass of mirror, MG , is given by

$$M_G = \vec{oT} \times \vec{N} \times (-F_z) = [0, r_n m g \sin \theta, 0]. \quad (20)$$

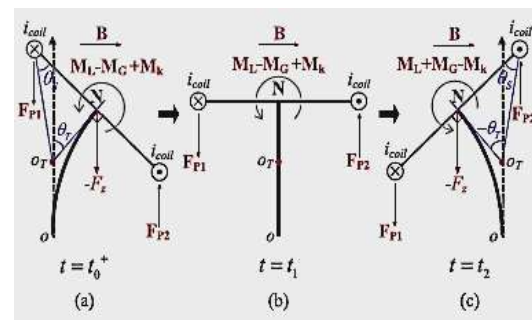


Fig. 12. Movement of the active mirror from (a) $t_0 +$ to (b) t_1 and to (c) t_2 .

where rN is the distance between points oT and N . In addition, the moment at point oT by spring resilience, Mk , can be expressed as $Mk = [0, kb\theta, 0]$. (21)

Since all of the moments consist only of y-axis components, the net scalar moment M_{net} at point oT is the sum of all of the scalar moments applied at oT ,

$$M_{net} = wFL \cos \theta + r_n m g \sin \theta - kb\theta. \quad (22)$$

M_{net} is also the product of the scalar moment of inertia I_{md} and angular acceleration, which can be expressed as

$$M_{net} = I_m \frac{d^2 \theta}{dt^2} = wFL \cos \theta + r_n m g \sin \theta - kb\theta \quad (23)$$

However, Eq. (23) is the second-order nonlinear ordinary differential equation without an analytical solution. Assuming that $\sin \theta \approx \theta$ and $\cos \theta \approx 1$ for small θ , it can be rewritten as

$$I_m \frac{d^2 \theta}{dt^2} = wFL + r_n m g \theta - kb\theta. \quad (24)$$

This can also be rewritten as

$$\frac{d^2\theta}{dt^2} + p\theta = q, \quad (25)$$

where $p = (kb - rNm_g)/I_m$ and $q = wFL/I_m$. From the initial conditions $\theta(0)=0$ and $d\theta/dt|_{t=0} = 0$, the solution for Eq. (25) can be obtained as

$$\theta(t) = \frac{q}{p}[1 - \cos(\sqrt{p}t)]. \quad (26)$$

Thus, time t_{in} phase 0 is dependent on angle θ and can be represented as

$$t(\theta) = \frac{1}{\sqrt{p}} \cos^{-1} \left(1 - \frac{p\theta}{q} \right). \quad (27)$$

By setting $\theta = \theta_T$, the time at the target tilting in phase 0, t_0 , can be obtained as

$$t_0 = t(\theta_T) = \frac{1}{\sqrt{p}} \cos^{-1} \left(1 - \frac{p\theta_T}{q} \right) \quad (28)$$

2) Phase 1: Once the direction of the current is reversed at t_0 , the Lorentz force is applied to the opposite direction for the range of $t_0 \leq t \leq t_1$, as shown in Fig. 12. Then, the tilting of the mirror plane is in a counter-clockwise direction. Thus, the net scalar moment M_{net} at point O_T can be expressed as

$$M_{net} = -wFL \cos\theta + r_n m g \sin\theta - k_b \theta. \quad (29)$$

For a small θ , similar to Eq. (4), it can be approximated that

$$I_m \frac{d^2\theta}{dt^2} = -wFL + r_n m g \theta - k_b \theta. \quad (30)$$

It can also be rewritten as

$$\frac{d^2\theta}{dt^2} + p\theta = -q. \quad (31)$$

From the boundary conditions $\theta(t_0) = \theta_T$ and $d\theta/dt|_{t=t_0} = 0$, the solution for Eq. (31) can be obtained as

$$\theta(t) = \frac{(q/p + \theta_T) \cos(\sqrt{p}(t - t_0)) - q/p}{1}. \quad (32)$$

The solution $t = t_1$ for making $\theta(t_1) = 0$ can be obtained as

$$t_1 = t_0 + \frac{1}{\sqrt{p}} \cos^{-1} \left(\frac{q/p}{q/p + \theta_T} \right). \quad (33)$$

3) Phase 2: The mirror plane is tilted counter-clockwise through the range of $t_1 \leq t \leq t_2$ during phase 2, as shown in Fig. 12, which leads to the net scalar moment equation M_{net} at point O_T becomes

$$M_{net} = -wFL \cos\theta - r_n m g \sin\theta + k_b \theta. \quad (34)$$

For a small θ , similar to Eq. (24), it can be approximated that

$$I_m \frac{d^2\theta}{dt^2} = -wFL - r_n m g \theta + k_b \theta. \quad (35)$$

It can also be rewritten as

$$\frac{d^2\theta}{dt^2} - p\theta = -q. \quad (36)$$

In order to obtain the boundary conditions at $t = t_1$, it is required to find $d\theta/dt|_{t=t_1}$ from Eq. (32) as

$$\left. \frac{d\theta}{dt} \right|_{t=t_1} = -\sqrt{p} \frac{(q/p + \theta_T)}{1} \sin(\sqrt{p}(t - t_0)) = v_{t1}. \quad (37)$$

By using the obtained boundary conditions $d\theta/dt|_{t=t_1} = v_{t1}$ and $\theta(t_1) = 0$, the solution for Eq. (36) can be obtained as

$$\theta(t) = \left[\frac{q}{p} \cos(\sqrt{p}(t - t_1)) - 1 \right] + \frac{v_{t1}}{\sqrt{p}} \sin(\sqrt{p}(t - t_1)) \quad (38)$$

By setting $t = t_2$ and $\theta(t_2) = -\theta_T$, t_2 can be obtained.

According to Eq. (16), the target tilting angle $\pm\theta_T$ can be controlled by adjusting times t_0 and t_2 . Once time t_0 is specified for target angle θ_T from Eq. (27), then the time $t_2 - t_0$ for keeping the tilting angle at $\pm\theta_T$ can also be determined from Eqs. (33) and (38). Note that the emission angle of the proposed on-road projection can simply be controlled by adjusting the timing for switching the direction of the current through the coil. Fig. 13 shows the emission angle by the law of reflection. Since the incidence angle is equal to the reflection angle, if the active mirror rapidly tilts from $-\theta_T$ to $+\theta_T$, the emission angle of an on-road projection covers $\pm 2\theta_T$. Thus, the emission angle θ_E of the proposed on-road projection can be defined as

$$|\theta_E| = |2\theta_T|. \quad (39)$$

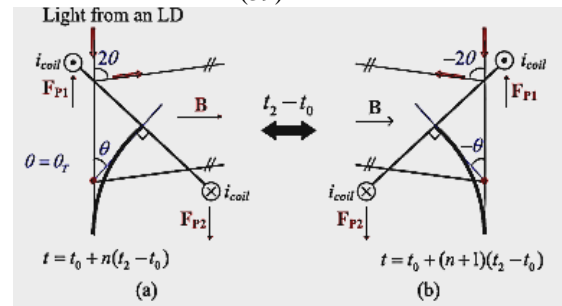


Fig. 13. Projected angle depending on the law on the law of reflection at the time (a) $t_0 + n(t_2 - t_0)$ and (b) $t_0 + (n + 1)(t_2 - t_0)$.

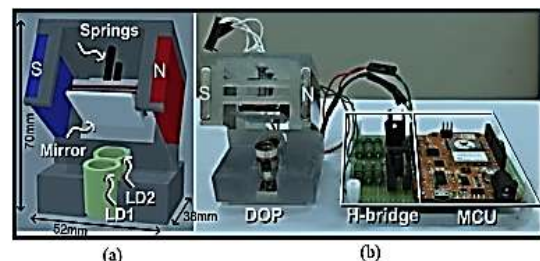


Fig. 14. (a) 3D rendering of the proposed active mirror and (b) the implementation of the proposed dynamic on-road projection system.

on-road projection covers $\pm 20T$. Thus, the emission angle θE of the proposed on-road projection can be defined as

$$|\theta E| = |20T| \quad (39)$$

IV. SIMULATION AND EXPERIMENTAL RESULTS

The 3D rendering of the proposed active mirror for the on-road projection technique is shown in Fig. 14(a). The two springs connecting the active mirror with the main base are used to allow the active mirror to tilt without pitch and yaw. Table II shows the design parameters used to obtain the center of the tilting axis, oT , as discussed in the lateral buckling equations for the spring. From the calculations in Eqs. (10) and (15), the length loT from the origin o to oT is 6.17mm. The design parameters for the dynamic analysis are shown in Table II as well. The prototype of the proposed dynamic on-road projection system is shown in Fig. 14(b). Fig. 15 shows the experimental results of the H-bridge in the proposed system. For the current controlling voltage $V_{REF} = 0.6$ V, the maximum value of i_{coil} is regulated to 200mA at $R_{HB} = 3 \Omega$. The switching frequency is purely controlled by the input HB , where $f_{HB} = 7.5$ Hz in this case. The comparisons between the simulation and experimental results for i_{coil} depending on V_{REF} are shown in Fig. 16. For $R_{HB} = 2.99 \Omega$, the maximum error from the simulation results is less than 2.17%.

Fig. 17 compares the simulation and experimental results of the emission angle θE as adjusted by both f_{HB} and i_{coil} based on Eqs. (16), (27), (33), (38), and (39). θE over f_{HB} is represented as a nonlinear graph, since θE is a sinusoidal function from Eq.

TABLE II
DESIGN PARAMETER FOR PROTOTYPE OF THE PROPOSED SYSTEM

Symbol	QUANTITY	Value
k	spring constant	154.35 N/m
l_s	length of spring	15 mm
m	mass of mirror plane	5 g
g	acceleration of gravity	9.81 m/s ²
E	elastic modulus	0.24 MPa
I_m^*	2nd moment of inertia of 2 springs	27.4 pm ⁴
l_{oT}	length of line from o to oT	6.17 mm
n	turns of coil	45
l	wire length affected by Lorentz force	8 mm
w	width of the mirror	24 mm
d	thickness of mirror plane	9 mm
r_N	length between points N and oT	13.33 mm
k_s	lateral bending stiffness of 2 springs	1050 $\mu N \cdot m/rad$
I_m^{**}	moment of inertia of mirror plane	0.99 $\mu kg \cdot m^2$

* I_m^* is approximated as a type of a rectangular hollow section.

** I_m^{**} is approximated as a type of a annular cylinder about axis [30].

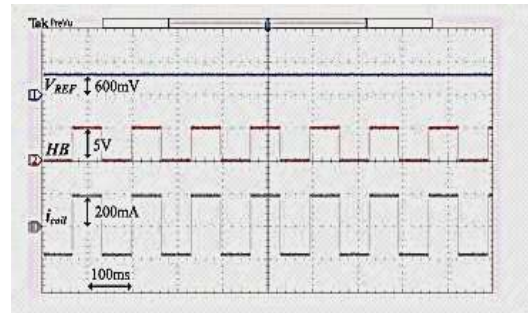


Fig. 15. Experimental results of the H-bridge at $f_{HB} = 7.5$ Hz.

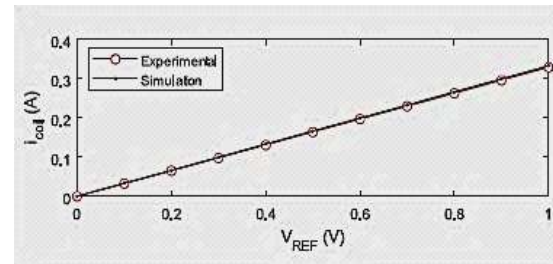


Fig. 16. Experimental results of H-bridge circuit for V_{REF} vs. i_{coil} .

(26) and the switching frequency $f_{HB} = 0.5(t_1 - t_2) - 1$ depends on t_0 . θE for the ranges of $5.75 \text{ Hz} \leq f_{HB} \leq 8.50 \text{ Hz}$ and $70 \text{ mA} \leq i_{coil} \leq 200 \text{ mA}$ are given by $4.28^\circ \leq |\theta E| \leq 38.72^\circ$ and $3.75^\circ \leq |\theta E| \leq 60.51^\circ$, respectively. The mismatch of θE between the simulation and experimental results is caused by several reasons: First, in order to determine the solution of θ in terms of time in Eqs. (24), (30), it is assumed that the tilting angle of the mirror is small. Therefore, the larger the tilting angle of the mirror, the larger the error that is introduced. Secondly, the magnetic flux density B within two magnets is assumed to be a

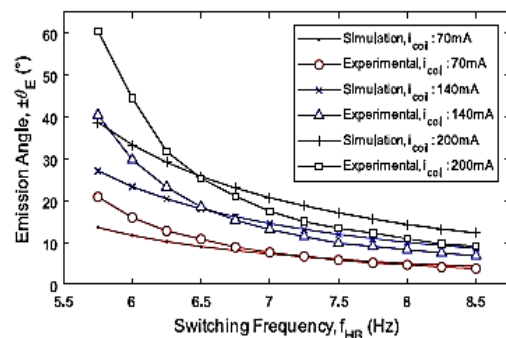


Fig. 17. Comparison between the simulation and experimental results for emission angle θE depending on various values of f_{HB} and i_{coil} at $B = 95 \text{ mT}$.

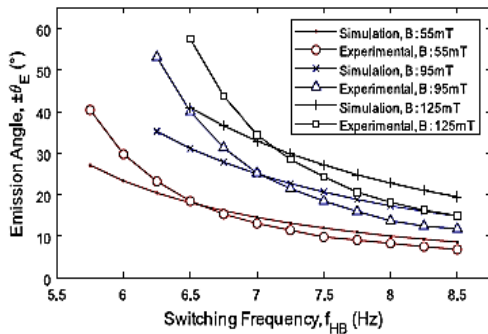


Fig. 18. Comparison between the simulation and experimental results for emission angle θ_E depending on various values of f_{HB} and B at icoil = 140 mA.

constant, but it depends on the location, which can be defined as

$$B = \mu_0 mD / 2\pi r^3 \quad (40)$$

where μ_0 is a constant known as the permeability of free space, mD is the magnetic moment of a magnetic dipole, and r is the distance from the magnetic dipole [31]. When the active mirror tilts in one direction, one side of the mirror moves close to a magnet while the other side moves away from the other magnet. Thus, the side of the active mirror closer to the magnet is influenced by a stronger magnetic flux than the other side, which leads to the total moment by Lorentz force becoming larger as the tilting angle increases. Finally, the non-linearity of the spring constant leads to errors. The spring usually has a different spring constant depending on the angle of tilting. Fig. 18 compares the simulation and experimental results of emission angle θ_E adjusted by both f_{HB} and B . Similar to Fig. 17, error between the simulation and experimental results is introduced. The error can be reduced by using a lookup table or a compensation method. Fig. 19 shows the top view of various patterns of the proposed dynamic on-road projection. The prototype is mounted 80 cm above the ground in order to mimic the position of a tail light. The on-road projected patterns are 2.5 m and 4.1 m tall for laser diodes 1 (LD1) and 2 (LD2), respectively. The emission angle

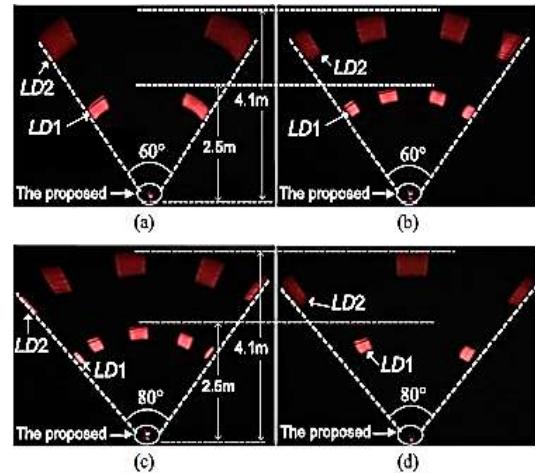


Fig. 19. On-road projection with various dynamic patterns by controlling the inputs f_{LD1} , D_{LD1} , f_{LD2} , D_{LD2} and f_{HB} . (a) $f_{LD1} = 12$ Hz, $D_{LD1} = 0.5$, $f_{LD2} = 12$ Hz, $D_{LD2} = 0.5$, $f_{HB} = 6$ Hz, (b) $f_{LD1} = 36$ Hz, $D_{LD1} = 0.5$, $f_{LD2} = 36$ Hz, $D_{LD2} = 0.5$, $f_{HB} = 6$ Hz, (c) $f_{LD1} = 46$ Hz, $D_{LD1} = 0.4$, $f_{LD2} = 46$ Hz, $D_{LD2} = 0.4$, $f_{HB} = 5.75$ Hz, (d) $f_{LD1} = 15$ Hz, $D_{LD1} = 0.15$, $f_{LD2} = 23$ Hz, $D_{LD2} = 0.25$, $f_{HB} = 5.75$ Hz.

With the range of $3.75^\circ \leq |\theta_E| \leq 60.51^\circ$ generates patterns with widths from 0.53m to 7.14m, which easily covers the full width of any vehicle. Fig. 19(a) shows the dynamic pattern with $|\theta_E| = 30^\circ$ at $f_{LD1} = 12$ Hz, $D_{LD1} = 0.5$, $f_{LD2} = 12$ Hz, $D_{LD2} = 0.5$, and $f_{HB} = 6$ Hz. If the switching frequency of each laser diode is increased while the rest of the parameters are unchanged, the projected pattern can be divided into more pieces, as shown in Fig. 19(b), where $f_{LD1} = 36$ Hz and $f_{LD2} = 36$ Hz. As shown in Fig. 19(c), the switching frequency of each laser diode is further increased to $f_{LD1} = f_{LD2} = 46$ Hz in order to make more pieces in the pattern. The duty-cycle of both laser diodes, $D_{LD1} = D_{LD2} = 0.4$, determines the ratio of light and dark patterns. $|\theta_E| = 40^\circ$ is achieved by setting $f_{HB} = 5.75$ Hz. Different patterns for LD1 and LD2 can also be formed without changing the emission angle by adjusting $f_{LD1} = 15$ Hz, $D_{LD1} = 0.15$, $f_{LD2} = 23$ Hz, and $D_{LD2} = 0.25$, as shown in Fig. 19(d). The specifications of the proposed system are as follows: $V_{DD} = 3.5$ V, icoil = 140 mA, and $B = 55$ mT. The two VCSEL diodes have the specifications of $V_{LD} = 2.2$ V, $i_{LD1} = 18$ mA, and $i_{LD2} = 18$ mA. The horizontal and vertical emission angles are ± 2.86 degree and ± 4.76 degree, respectively. The power consumption of the proposed system, excluding the MCU, is 0.98W. Comparisons with other possible on-road projections in Table III show that the proposed technique is superior to others by

generating a dynamic pattern in wide angle without requiring expensive and bulky customized optics.

V. DYNAMIC ON-ROAD PROJECTION WITH A ULTRASONIC SENSOR

The dynamic pattern of the proposed on-road projection is controlled in conjunction with a conventional ultrasonic sensor

TABLE III
 COMPARISONS WITH OTHER TECHNIQUES

	COLLIMATING LENS [11]	DMDs [13]	This work
Dynamic pattern	X	O	O
Customized Optics-free	X	X	O
Projected Angle	$\pm 20^\circ$	$\pm 39^\circ$	$\pm 70^\circ$

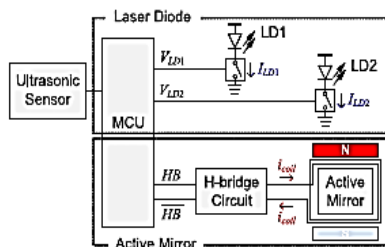


Fig. 20. Block diagram of the proposed on-road projection system with an ultrasonic sensor.

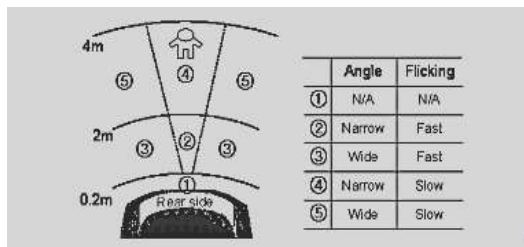


Fig. 21. Dynamic pattern control of the proposed on-road projection with an ultrasonic sensor based on the location of pedestrians.



Fig. 22. Projection angle control for LD1 and LD2. (a) Narrow and (b) wide.

In a vehicle as shown in Fig. 20. Depending on the location of pedestrians detected by the ultrasonic sensor, the projection angle and the flicking of laser patterns can be controlled as shown in Fig. 21. When pedestrians are too close to the rear

bump around region 1, dynamic on-road projection is disabled for eye safety. The projection angle covers the location of pedestrians detected by an ultrasonic sensor. In order to warn pedestrian as the vehicle gets closer, the flicking of laser pattern becomes faster. Since an ultrasonic sensor can detect multiple points, the proposed system corresponds to the multiple pedestrians. Fig. 22 shows the two different angular control of the projected laser pattern without adjusting the duty-cycle to verify the maximum



Fig. 23. Implementation of the proposed on-road projection with an ultrasonic sensor.

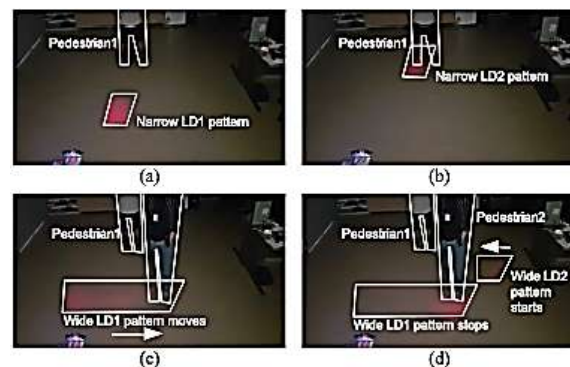


Fig. 24. Projected on-road real-time dynamic pattern with two LDs depending on the location of pedestrians: (a)(b) Pedestrian located 3m away from the sensor (pattern: narrow and slow) (c)(d) Pedestrian located 1.5m away from the sensor (pattern: wide and fast).

Projected angle of $\pm 70^\circ$ at the maximum power of 0.98W, where $V_{DD} = 4.5\text{ V}$, $i_{coil} = 200\text{ mA}$, $V_{LD} = V_{LD2} = 2.2\text{ V}$, $i_{LD1} = i_{LD2} = 18\text{ mA}$, and $B = 55\text{ mT}$. The prototype is implemented as depicted in Fig. 23. Even though there are two MCUs for the prototype, one for the ultrasonic sensor and the other for projection control, it can be integrated in a single MCU. Fig. 24 shows the projected real-time dynamic pattern corresponds to the control scheme in Fig. 21 depending on the location of pedestrians [32].

VI. CONCLUSION

In order to avoid collisions between vehicles and pedestrians, vehicle warning systems have been widely researched. This paper presents a vehicle warning system using dynamic patterns projected onto the road with a wide angle. In order to generate various dynamic patterns, an active mirror and laser diodes are controlled in terms of frequency and duty-cycle. The proposed system emits various dynamic patterns by using two laser diodes with emission angles up to $\pm 70^\circ$. The proposed technique along with an ultrasonic sensor technology for pedestrian recognition generates dynamic patterns automatically projected depending on the position of pedestrians. In addition, the proposed active mirror in conjunction with a micro-electro mechanical system (MEMS) technology can be implemented in an area much smaller than the prototype.

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